

ELEC0014 - Introduction to electric power and energy systems

The power transformer

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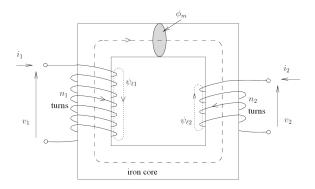
November 2019

Power transformers are used:

- to transmit electrical energy under high voltages
 - step-up transformers at the terminal of generators
 - step-down transformers to distribute energy at the end-users
- to control the voltages at some busbars:
 - in sub-transmission networks
 - in distribution networks
- to control the power flows in some parts of a meshed network.

The single-phase transformer

Principle



step-up transformer: secondary voltage > primary voltage step-down transformer: secondary voltage < primary voltage.

alternating voltage v_1 at terminals of coil 1 \longrightarrow alternating current i_1 in coil 1 \longrightarrow alternating magnetic field \longrightarrow voltage induced in coil 2 \longrightarrow current i_2 in coil 2 \longrightarrow magnetic field superposed to the one created by i_1 .

Fluxes in coils

$$n_1i_1+n_2i_2=\mathcal{R}\phi_m$$

 ϕ_m : magnetic flux in a cross section of the iron core \mathcal{R} : *reluctance* of the magnetic circuit

+ sign: due to the way coils are wound and the direction of currents

$$\psi_1 = \psi_{\ell 1} + n_1 \phi_m$$

 ψ_1 : flux linkage in coil 1

 $\psi_{\ell 1}$: *leakage flux* in coil 1 (lines of magnetic field crossing coil 1 but not passing through the iron core)

$$\psi_2 = \psi_{\ell 2} + n_2 \phi_m$$

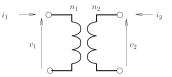
- ψ_2 : flux linkage in coil 2
- $\psi_{\ell 2}$: *leakage flux* in coil 2 (lines of magnetic field crossing coil 2 but not passing through the iron core)

Ideal transformer

- The coils have no resistance
- the coils have no leakage flux
- the permeability of the core material is infinite.

$$\mathcal{R}\simeq 0 \quad \Rightarrow \quad i_2=-\frac{n_1}{n_2}i_1$$

$$v_1 = \frac{d\psi_1}{dt} = n_1 \frac{d\phi_m}{dt} \quad v_2 = \frac{d\psi_2}{dt} = n_2 \frac{d\phi_m}{dt} \quad \Rightarrow v_2 = \frac{n_2}{n_1} v_1$$



 $v_1 i_1 = -v_2 i_2$: no losses in the ideal transformer!

Equations of the real transformer

Leakage inductances: $L_{\ell 1} = \frac{\psi_{\ell 1}}{i_1}$ $L_{\ell 2} = \frac{\psi_{\ell 2}}{i_2}$ Magnetizing inductance (seen from coil 1) : $L_{m1} = \frac{n_1^2}{\mathcal{R}}$

Flux linkages:

$$\psi_1 = L_{\ell 1} \, \dot{i}_1 + n_1 \frac{n_1 \dot{i}_1 + n_2 \dot{i}_2}{\mathcal{R}} = L_{\ell 1} \, \dot{i}_1 + \frac{n_1^2}{\mathcal{R}} \dot{i}_1 + \frac{n_1 n_2}{\mathcal{R}} \dot{i}_2 = L_{\ell 1} \, \dot{i}_1 + L_{m 1} \dot{i}_1 + \frac{n_2}{n_1} L_{m 1} \dot{i}_2$$

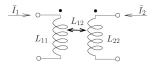
$$\psi_2 = L_{\ell 2} i_2 + n_2 \frac{n_1 i_1 + n_2 i_2}{\mathcal{R}} = L_{\ell 2} i_2 + \frac{n_2^2}{\mathcal{R}} i_2 + \frac{n_1 n_2}{\mathcal{R}} i_1 = L_{\ell 2} i_2 + (\frac{n_2}{n_1})^2 L_{m 1} i_2 + \frac{n_2}{n_1} L_{m 1} i_1$$

Voltages at the terminals of the coils:

$$v_{1} = R_{1}i_{1} + \frac{d\psi_{1}}{dt} = R_{1}i_{1} + L_{\ell 1}\frac{di_{1}}{dt} + L_{m 1}\frac{di_{1}}{dt} + \frac{n_{2}}{n_{1}}L_{m 1}\frac{di_{2}}{dt}$$

$$v_{2} = R_{2}i_{2} + \frac{d\psi_{2}}{dt} = R_{2}i_{2} + L_{\ell 2}\frac{di_{2}}{dt} + (\frac{n_{2}}{n_{1}})^{2}L_{m 1}\frac{di_{2}}{dt} + \frac{n_{2}}{n_{1}}L_{m 1}\frac{di_{1}}{dt}$$

The transformer is a particular case of magnetically coupled circuits

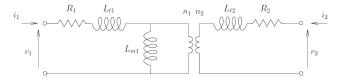


- if the currents enter by the terminals marked with $\bullet,$ their contributions to the flux ϕ_m are added
- if the currents are counted positive when entering the terminals marked with •, the mutual inductance is positive
- the ${\scriptstyle \bullet}$ also indicate AC voltages which are in phase when the transformer is supposed ideal.

One easily identifies:

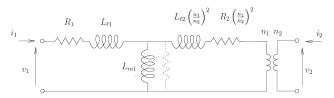
$$L_{11} = L_{\ell 1} + \frac{n_1^2}{\mathcal{R}}$$
 $L_{12} = \frac{n_1 n_2}{\mathcal{R}}$ $L_{22} = L_{\ell 2} + \frac{n_2^2}{\mathcal{R}}$

Equivalent circuits of the real transformer



 $R_1i_1^2 + R_2i_2^2$: copper losses

Passing R_2 et $L_{\ell 2}$ from side 2 to side 1:

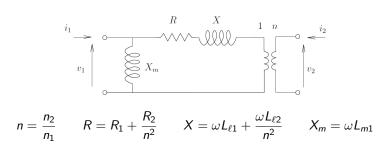


Possible improvements:

- shunt resistance to account for iron losses (due to eddy currents)
 - kept small by using laminated cores
 - negligible compared to the power passing through the transformer
- non-linear inductance L_{m1} to account for iron saturation.

Usual simplified equivalent circuit

Usual simplification taking into account that:



 $\omega L_{m1} >> R_1, R_2, \omega L_{\ell 1}, \omega L_{\ell 2}$

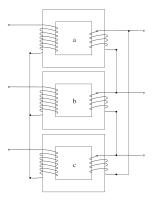
• Equivalent circuit justified by the measurements provided by manufacturers

- X = leakage reactance (combined)
- X = short-circuit reactance = reactance seen from port 1 when port 2 is short-circuited (considering that X_m is very large compared to X)

The three-phase transformer

First type

Three separate single-phase transformers. No magnetic coupling between phases.

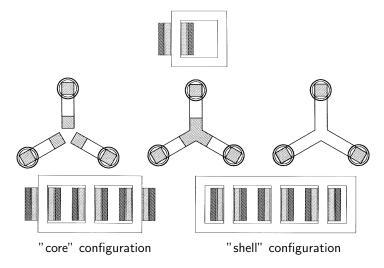


Appropriate for transformers of large nominal power:

- in case of failure of one of the transformers, only that transformer is replaced
- easier to carry.

Second type

The three phases are mounted on a common iron core.



Volume of the common core smaller than three times the volume of a single core.

Star vs. Delta configuration

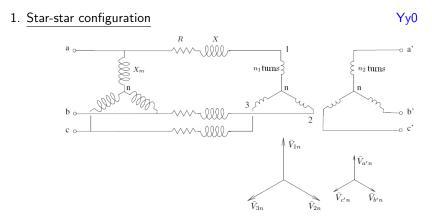
Four possible mountings.

- Transformer side connected to extra high voltage network: star configuration preferred since the voltage across each winding is $\sqrt{3}$ times smaller
- star configuration: possibility to connect the neutral to ground (safety)
- star configuration preferred to place a load tap changer (see later in this chapter)
- transformer side carrying high currents (f.i. the generator side of a step-up transformer): delta configuration preferred since the currents in the branches are $\sqrt{3}$ times smaller
- delta configuration used to eliminate the harmonics of order 3, 6, 9, etc.

The power transformer The three-phase transformer

Single-phase equivalents of three-phase transformers

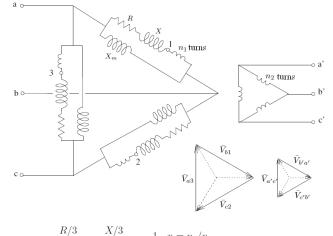
- transformer with single core: the phases are magnetically coupled \rightarrow perform a per-phase analysis (see Chapter 2)
- for simplicity, we consider a transformer with three separate cores
- we focus on the impact of the star vs. delta configuration.

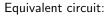


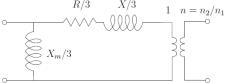
Per-phase equivalent circuit = equivalent circuit of one phase.

2. Delta-delta configuration



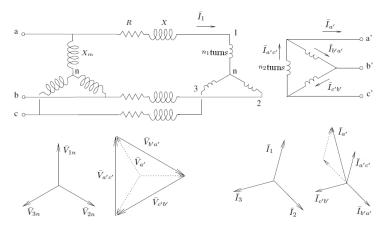






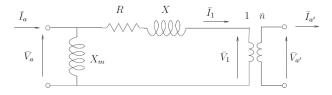
3. Star-delta configuration





$$\bar{V}_{a'} = \frac{1}{\sqrt{3}} e^{j\pi/6} \bar{V}_{a'c'} = \frac{n_2}{\sqrt{3}n_1} e^{j\pi/6} \bar{V}_{1n} = \bar{n}\bar{V}_{1n} \quad \text{où} \quad \bar{n} = \frac{n_2}{\sqrt{3}n_1} e^{j\pi/6}$$
$$\bar{I}_{a'} = \bar{I}_{a'c'} - \bar{I}_{b'a'} = \sqrt{3} e^{j\pi/6} \bar{I}_{a'c'} = \frac{\sqrt{3}n_1}{n_2} \frac{1}{e^{-j\pi/6}} \bar{I}_1 = \frac{1}{\bar{n}^*} \bar{I}_1$$

Equivalent circuit:



Ideal transformer with complex ratio \bar{n} :

- is characterized by : $\bar{V}_{a'} = \bar{n}\bar{V}_1$ $\bar{I}_{a'} = \bar{I}_1/\bar{n}^{\star}$
- reduces to the standard ideal transformer if \bar{n} is real
- transfers complex power without losses:

$$\bar{V}_{a'} \ \bar{I}_{a'}^{\star} = \bar{n} \ \bar{V}_1 \ \frac{1}{\bar{n}} \ \bar{I}_1^{\star} = \bar{V}_1 \ \bar{I}_1^{\star}$$

The above two-port is non reciprocal:

$$[\bar{I}_a]_{\bar{V}_a=0,\bar{V}_{a'}=1} \neq -[\bar{I}_{a'}]_{\bar{V}_a=1,\bar{V}_{a'}=0}$$

4. Delta-star configuration

Derivation similar to that of the Star-delta configuration,

leading to a single-phase equivalent circuit with:

• the complex transformer ratio:

$$ar{n} = rac{\sqrt{3} n_2}{n_1} e^{-j\pi/6}$$

- a series resistance R/3
- a series reactance X/3
- a shunt reactance $X_m/3$.

Designation of a transformer

Standardized abbreviation of I.E.C. (International Electrotechnical Commission)

Also referred to as vector group of a transformer

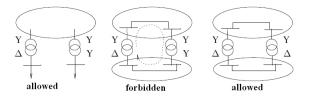
3 symbols:

- $\bullet\,$ an uppercase letter for the high-voltage side: ${\bf Y}$ for a star connection or ${\bf D}$ for a delta
- ullet a lowercase letter for the low-voltage side: ${\bf y}$ for a star connection or ${\bf d}$ for a delta
- an integer $p \in \{0, 1, \dots, 11\}$:
 - an indication of the phase displacement between the primary and secondary voltages of the same phase, the transformer being assumed ideal
 - the phasor of the high voltage being on the number 12 of a clock, *p* is the number pointed by the phasor of the low voltage

and for the star configuration:

 \boldsymbol{n} after \boldsymbol{y} or \boldsymbol{Y} to indicate that the neutral is grounded.

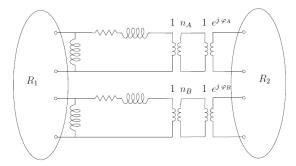
Caution as regards using transformers with different phase displacements



When a given sub-network is fed by two (or more) transformers operating "in parallel" (i.e. located in at least one loop), the latter must have the same phase displacement p.

Otherwise, the different phase displacements would cause unacceptable power flows.

Simplification of computations



Two transformers with the same phase displacement: $\varphi_A = \varphi_B = \varphi$

The ideal transformers with complex ratio $e^{j\varphi}$ can be removed without changing:

- the magnitudes of the branch currents and bus voltages
- the complex powers flowing in the branches.

The phase displacements in the transformer models are ignored when computing the steady-state balanced operation of power systems.

Nominal values, per unit system and orders of magnitudes

Nominal values

• Nominal primary voltage U_{1N} and nominal secondary voltage U_{2N} : voltages for which the transformer has been designed (in particular its insulation).

The real voltages may deviate from these values by a few %.

• Nominal primary current I_{1N} and nominal secondary current I_{2N} : currents for which the transformer has been designed (in particular the cross-sections of the conductors).

Maximum currents that can be accepted without limit in time.

• nominal apparent power S_N :

$$S_N = \sqrt{3} U_{1N} I_{1N} = \sqrt{3} U_{2N} I_{2N}$$

Conversion of parameters in per unit values

- choose the (three-phase) base power $S_B = S_N$
- on primary side, choose the (phase-to-neutral) base voltage $V_{1B} = U_{1B}/\sqrt{3}$
- on secondary side, choose the (phase-to-neutral) base voltage $V_{2B}=U_{2B}/\sqrt{3}$
- the impedances of the equivalent circuit, which are located on the primary side, are divided by $Z_{1B} = 3V_{1B}^2/S_B = U_{1B}^2/S_B$
- the value of the transformer ratio $n = n_2/n_1$ in per unit is obtained as follows:

$$v_{2} = \frac{n_{2}}{n_{1}}v_{1} \quad \Leftrightarrow \quad v_{2pu} = \frac{v_{2}}{V_{2B}} = \frac{n_{2}}{n_{1}}\frac{v_{2}}{V_{2B}}v_{1} = \frac{n_{2}}{n_{1}}\frac{V_{1B}}{V_{2B}}\frac{v_{1}}{V_{1B}} = \frac{n_{2}}{n_{1}}\frac{V_{1B}}{V_{2B}}v_{1pu}$$
$$\Rightarrow \quad n_{pu} = \frac{n_{2}}{n_{1}}\frac{V_{1B}}{V_{2B}}$$

If $V_{2B}/V_{1B} = n_2/n_1$: $n_{pu} = 1$: the ideal transformer disappears from the equivalent circuit !

In practice, $V_{2B}/V_{1B} \simeq n_2/n_1$: the ideal transformer remains in the equivalent circuit but with a ratio $n_{pu} \simeq 1$.

Orders of magnitude

resistance R	< 0.005 pu
leakage reactance $^1~\omega L$	
magnetizing reactance ωL_m	range: 20 – 50 pu
transformer ratio $n = n_2/n_1$	range: 0.85 — 1.15 pu

values on the (S_B, V_{1B}, V_{2B}) base of the transformer !!

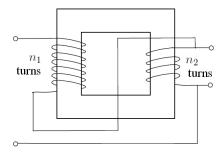
Network computation in another base: convert the parameters to that base (see formula in the chapter on per unit system)

¹or short-circuit reactance

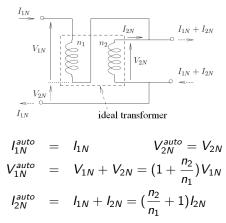
Autotransformers

Single-phase autotransformer

Transformer whose primary and secondary sides are connected in such a way that they have a winding in common:



Let us assume that the inner transformer operates with its voltages and currents at their nominal values (all losses neglected, transformer assumed ideal).



Ratio of the autotransformer ?

$$n^{auto} = \frac{V_{2N}^{auto}}{V_{1N}^{auto}} = \frac{V_{2N}}{V_{1N} + V_{2N}} = \frac{\frac{n_2}{n_1}}{1 + \frac{n_2}{n_1}}$$

For the chosen primary and secondary, the transformer is of the step-down type.

Nominal apparent power of the autotransformer ?

$$S_N^{auto} = V_{1N}^{auto} I_{1N}^{auto} = (1 + \frac{n_2}{n_1}) V_{1N} I_{1N} = (1 + \frac{n_2}{n_1}) S_N$$

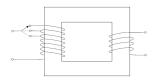
- The autotransformer allows for a power transfer higher than S_N . \Rightarrow reduced investment costs and reduced losses !
- True for any n_1, n_2 values but for a higher "amplification": $n_2 \gg n_1$
- However, if $n_2 \gg n_1$, the autotransformer ratio $n^{auto} \rightarrow 1$. Hence, the device cannot connect two very different voltage levels
- Autotransformers used to transfer high powers between two networks with relatively close nominal voltages
 - $\bullet\,$ Belgium : 550 MVA autotransformers between 400 and 150 kV
 - France: autotransformers between 400 and 225 kV.
- drawback: metallic connection between primary and secondary ⇒ voltage disturbances propagate more easily.

Three-phase autotransformer

Assembly of three single-phase autotransformers.

Adjustment of the turn ratio





- Objective: adjust voltage at a busbar (usually one of the transformer ends)
- adjustment in steps: between 15 and 25 tap positions
- to modify the number of turns in service:
 - transformer taken out of service
 - transformer kept in service: the on-load (or under-load or load) tap changer modifies the windings without interrupting the current (avoid electric arcs!)
- load tap changers can be controlled
 - manually: remotely by operator supervising the network from a control center
 - automatically: local feedback system (see chapter on voltage control)
- placement of tap changer:
 - usually on the high-voltage side: current smaller, more turns in winding
 - three-phase transformer: near neutral in Y configuration (lower voltages).

Accounting the tap position changes in equivalent circuit

In principle, one set of $(R, \omega L, \omega L_{m1}, n)$ values for each tap position.

In practice, ωL and n are the most affected, while $R \ll$ and $\omega L_{m1} \gg$.

Possible simplification: let us assume that:

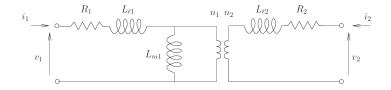
- the turns are adjusted on side 2 in equivalent circuit
- the leakage inductance $L_{\ell 2}$ vary with the number of turns n_2 according to:

$$L_{\ell 2} = L_{\ell 2}^{o} (\frac{n_2}{n_2^{o}})^2$$

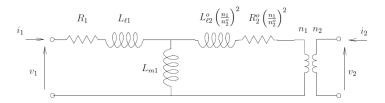
• and similarly for the resistance R₂:

$$R_2 = R_2^o (\frac{n_2}{n_2^o})^2$$

This is arguable, but R_2 is small...



After passing R_2 and $L_{\ell 2}$ on the other side of the ideal transformer:



When the tap position (and, hence, the number of turns n_2) changes:

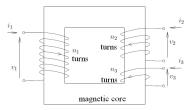
- impedances located on the non-adjusted side remain constant
- only the transformer ratio n_2/n_1 changes.

Three-winding transformers

Shortcut for "transformers with three windings per phase".

Principle

Single-phase transformer with 3 windings (= 1 phase of a 3-phase transformer) :

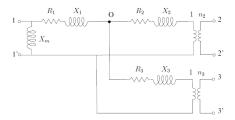


- Power transfer between three voltage levels
- share of power flows between the windings depends on what is connected to the transformer
- nominal apparent powers of the three windings usually different.

Other uses

- in switching stations, power supplied to auxiliaries by the third winding
- connection of a shunt inductance or capacitor for compensation purposes
- improvement of operation in unbalanced condition
- improvement of power quality in the presence of harmonics.

Equivalent circuit



 $R_1 + R_2 + j(X_1 + X_2)$: impedance seen from 1 with 2 short-circuited and 3 opened $R_1 + R_3 + j(X_1 + X_3)$: impedance seen from 1 with 3 short-circuited and 2 opened

Some reactances of this equivalent circuit can be negative (for instance if the windings have very different nominal apparent powers).

Phase shifting transformer

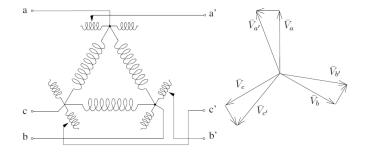
Also called simply phase shifter.

Transformer aimed at shifting the secondary voltage phasor with respect to the primary voltage phasor, in order to adjust active power flows in the network.

Two main configurations:

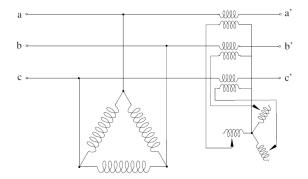
- transformer connecting two networks with different nominal voltages (as usual) to which a device is added to adjust the phase angle
- dedicated device, with the *same* primary and secondary nominal voltages, aimed at adjusting the phase angle.

First scheme



- adjustment in quadrature
- some variation of the voltage magnitude with the phase angle
- there exist more elaborate schemes where the voltage magnitude is kept constant while the phase angle is adjusted
- drawback of this scheme: the whole line current passes through the tap changer (unavoidable electric arcs).

Second scheme



- excitation shunt transformer + series transformer
- nominal voltage of series transformer = fraction of nominal phase-to-neutral voltage $V_N \Rightarrow$ nominal apparent power = fraction of $3V_N I_{max}$
- compared to previous scheme: lower current in the tap changer.

Example: phase shifting transformers on the borders of Belgium



380/380 kV : in series with:

- Iine Zandvliet (B) Borssele (NL) and Zandvliet (B) Geertruidenberg (NL)
- Iine Meerhout (B) Maasbracht (NL)
- Iine Gramme (B) Maasbracht (NL)
- nominal power $3V_N I_{max} = 1400$ MVA
- phase shift adjustment: 35 positions, $+17/-17 \times 1.5^{o}$ (at no load)

220/150 kV : in series with the Chooz (F) - Monceau (B) line

- nominal power: 400 MVA
- \bullet in-phase adjustment : 21 positions, +10/-10 \times 1.5 %
- quadrature adjustment: 21 positions, +10/-10 imes 1.2°